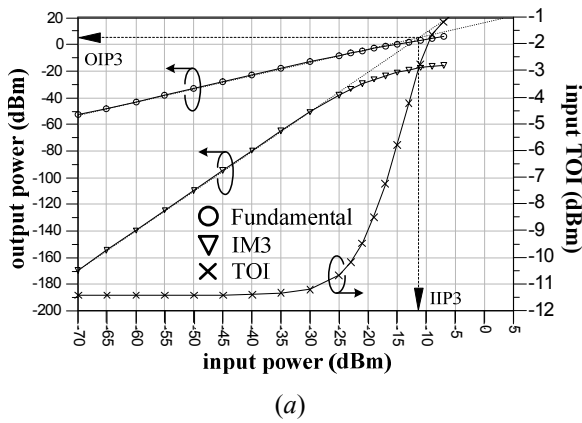
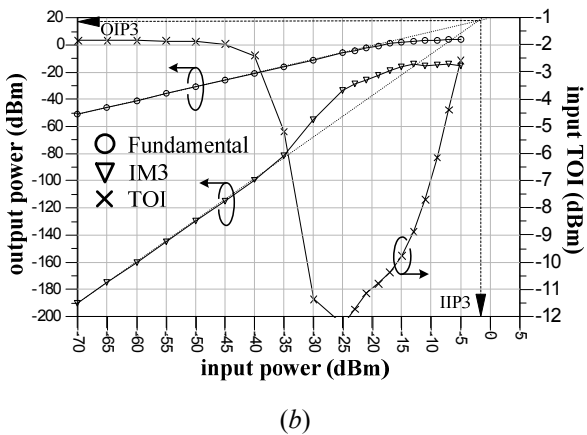


Fig. 5. LNA's simulated input and output return losses



(a)



(b)

Fig. 6. simulated linearity and input-referred intercept point for (a) HBT LNA and (b) pHEMT LNA

IV. CONCLUSION

A 2.4-GHz cascode LNA was designed and simulated in 0.2- μm GaAs Pseudomorphic HEMT process and 0.35- μm HBT SiGe BiCMOS process as a case of comparison between these two popular technologies. As illustrated, the pHEMT LNA consumes a DC current two times more than the HBT LNA to present the identical gain; but it gave the benefits of lower noise figure and wider matching band. Both LNAs presented an identical linearity, but the third-order intermodulation component grew sooner in pHEMT LNA than

the one in HBT LNA. As a result, the SiGe HBT process shows a perfect narrow-band characteristics and GaAs pHEMT is a proper choice for broadband and low-noise component designs. The HBT LNA and pHEMT LNA parameters are summarized in the table 1 to simplify the comparison.

TABLE I. PERFORMANCE SUMMARY

Parameters	Technology	
	GaAs pHEMT LNA	SiGe HBT LNA
Gain (dB)	16.6	19.2
NF (dB)	1.00	1.62
S ₁₁ (dB)	< -16.5	< -20.3
S ₂₂ (dB)	< -20.6	< -22.6
S ₁₂ (dB)	-22.5	-25
ICP _{1dB} (dBm)	-19	-17
IIP3 (dBm)	-11.4	-1.88
Current (mA)	28	14
Supply Voltage (V)	3	
Frequency (GHz)	2.4 ~ 2.5	
1-dB Gain BW (GHz)	> 3	> 1.5
Matching BW (GHz) [S ₁₁ < -10]	> 6.8	> 2.6
Matching BW (GHz) [S ₂₂ < -10]	> 3.4	> 1

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